Chapter 6 Random Processes and Spectral Analysis

**Matched filters** 



## 6-8 Matched filters



# A general representation for a matched filter is illustrated as follows:

6-8 Matched filters

$$r(t) = s(t) + n(t)$$

$$h(t)$$

$$h(t)$$

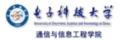
$$h(f)$$

$$r_0(t) = s_0(t) + n_0(t)$$

### Matched filters design criterion: To find *h*(*t*) or, equivalently, *H*(*f*), so that

$$\left(\frac{S}{N}\right)_{out} = \frac{s_0^2(t)}{n_0^2(t)}$$

is a maximum at a sampling time  $t=t_0$ .



## ~ 6-8 Matched filters General results r(t) = s(t) + n(t)**Matched filters** $r_0(t) = s_0(t) + n_0(t)$ h(t)H(f) $s_0(t_0) = \int H(f)s(f)e^{j\omega t_0}df$ $s_0(f) = H(f)s(f)$ $\overline{n_0^2(t)} = \int \left| H(f) \right|^2 p_n(f) df$ $\left(\frac{S}{N}\right)_{out} = \frac{s_0^2(t)}{n_0^2(t)} = \frac{\left|\int\limits_{-\infty}^{\infty} H(f)s(f)e^{j\omega t_0}df\right|^2}{\int\limits_{-\infty}^{\infty} |H(f)|^2 p_n(f)df}$

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### The matched filter is the linear filter that maximizes

$$\left(\frac{S}{N}\right)_{out} = \frac{s_0^2(t)}{\overline{n_0^2(t)}}$$

And that has a transfer function given by

$$H(f) = K \frac{S^*(f)}{p_n(f)} e^{-j\omega t_0}$$

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### **Theorem**

When the input noise is white, the impulse response of the matched filter becomes

$$h(t) = Cs(t_0 - t)$$

Where

*C* is an arbitrary real positive constant,  $t_0$  is the time of the peak signal output, s(t) is the known input–signal waveshape.



## 6-8 Matched filters Results for white noise An important property of matched filters:

the actual value of  $(S/M)_{out}$  can be obtained form the matched filter.

$$\left(\frac{S}{N}\right)_{out} = \int_{-\infty}^{\infty} \frac{\left|S(f)\right|^2}{p_n(f)} df$$
$$\left(\frac{S}{N}\right)_{out} = \int_{-\infty}^{\infty} \frac{\left|S(f)\right|^2}{N_0/2} df = \frac{2}{N_0} \int_{-\infty}^{\infty} \left|S(f)\right|^2 df = \frac{2}{N_0} \int_{-\infty}^{\infty} s^2(t) dt$$
$$\left(\frac{S}{N}\right)_{out} = \frac{2E_s}{N_0}$$

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The result states that  $(S/N)_{out}$  depends on the signal energy and PSD level of the noise, and not on the particular signal waveshape that is used. <sup>7</sup> Your site here

### (1) Integrate-and-dump matched filter

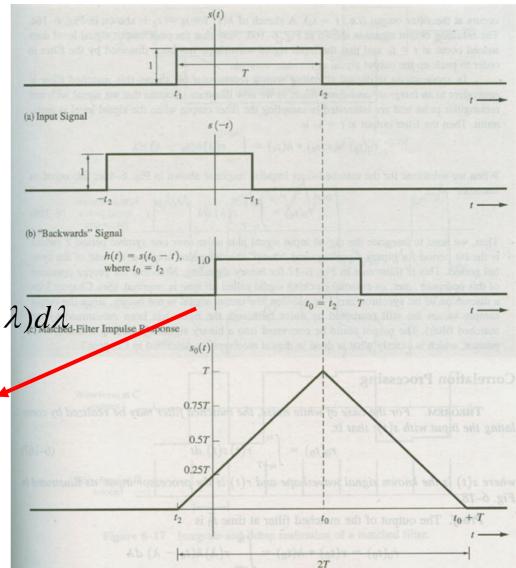
Example 6-11

$$s(t) = \begin{cases} 1 & , & t_1 \le t \le t_2 \\ 0 & , & others \end{cases}$$

$$h(t) = s(t_0 - t) = s(-(t - t_0))$$

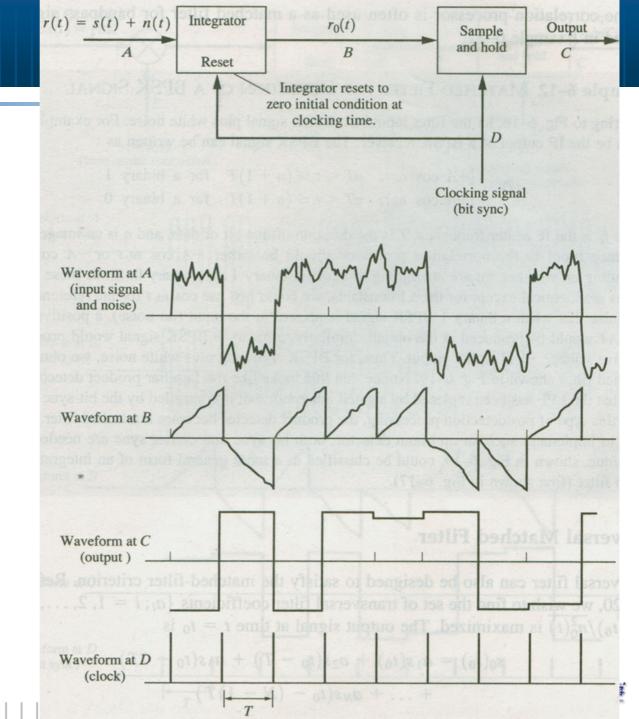
$$r_{0}(t_{0}) = r(t_{0}) * h(t_{0}) = \int_{-\infty}^{\infty} r(\lambda)h(t_{0} - t_{0}) = \int_{-\infty}^{\infty} r(\lambda)h(t_{0} - t_{0}) = \int_{0}^{\infty} r(\lambda)h(t_{0} - t$$

$$r_0(t_0) = \int_{t_0-T}^{t_0} r(\lambda) d\lambda$$

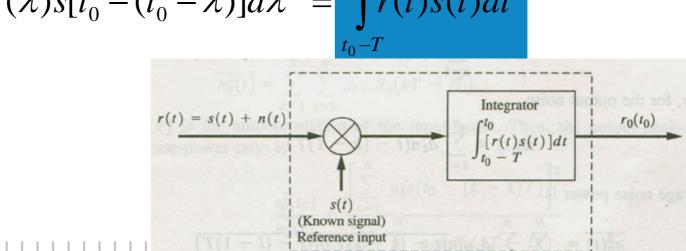


(1) Integrate-and-dump matched filter

Example 6-11



## Matched filter realization (2) Correlation processing For the case of white noise $r_0(t_0) = r(t_0) * h(t_0) = \int r(\lambda) h(t_0 - \lambda) d\lambda$ $h(t) = \begin{cases} s(t_0 - t) & 0 \le t \le T \\ 0 & others \end{cases}$ $r_0(t_0) = \int_0^{t_0} r(\lambda) s[t_0 - (t_0 - \lambda)] d\lambda = \int_0^{t_0} r(t) s(t) dt$





### (2) Correlation processing

Theorem:

For the case of white noise, the matched filter may be realized by correlating the input r(t) with s(t), that is,

$$r_0(t_0) = \int_{t_0-T}^{t_0} r(t)s(t)dt$$

Where s(t) is the known signal waveshape and r(t) is the processor input.



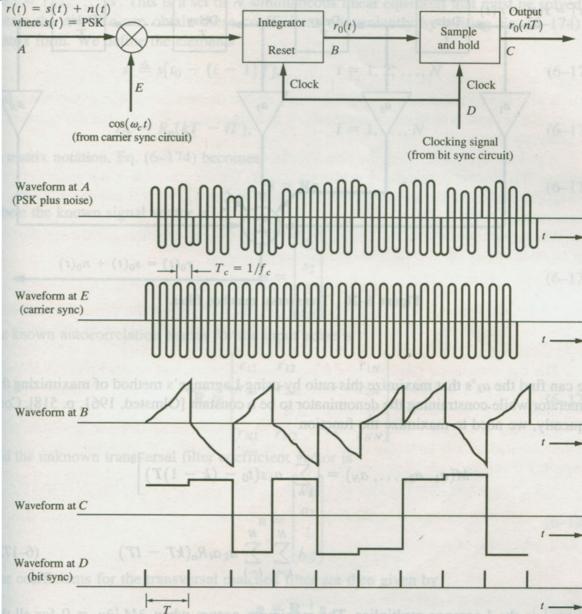
Example 6-12

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Matched filter for detection of a BPSK signal

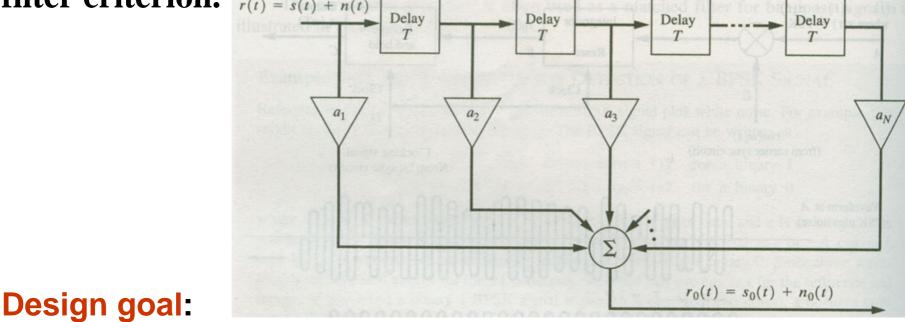
$$s(t) = \begin{cases} +A\cos\omega_c t & \text{for "1"} \\ -A\cos\omega_c t & \text{for "0"} \end{cases}$$
$$nT \le t \le (n+1)T$$

$$r_0(t_0) = \int_{t_0-T}^{t_0} r(t)s(t)dt$$



### (3) Transversal matched filter

### A transversal filter can be designed to satisfy the matchedfilter criterion. r(t) = s(t) + n(t)



To find the set of transversal filter coefficients  $\{a_i, i=1, 2, ..., N\}$  such that  $s_0^2(t_0)/\overline{n_0^2(t)}$  is maximized 13 Your site here

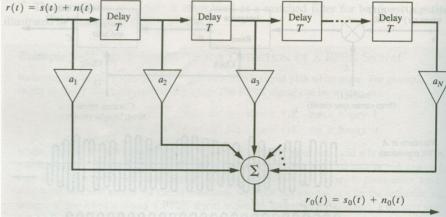
### (3) Transversal matched filter

 $s_0(t_0) = a_1 s(t_0) + a_2 s(t_0 - T) + a_3 s(t_0 - 2T) + \dots + a_N s[t_0 - (N - 1)T]$ 

$$= \sum_{k=1}^{N} a_k s(t_0 - (k-1)T)$$

$$n_0(t) = \sum_{k=1}^{N} a_k n(t - (k - 1)T)$$

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$$\overline{n_0^2(t)} = \sum_{k=1}^N \sum_{l=1}^N a_k a_l \ \overline{n(t - (k-1)T)} \ n(t - (l-1)T)$$

$$\frac{s_0^2(t_0)}{n_0^2(t)} = \frac{\left[\sum_{k=1}^N a_k s(t_0 - (k-1)T)\right]^2}{\sum_{k=1}^N \sum_{l=1}^N a_k a_l R_n(kT - lT)}$$



$$\frac{s_0^2(t_0)}{n_0^2(t)} = \frac{\left[\sum_{k=1}^N a_k s(t_0 - (k-1)T)\right]^2}{\sum_{k=1}^N \sum_{l=1}^N a_k a_l R_n(kT - lT)}$$

Using Lagrange's method of maximizing the numerator

$$M(a_{1}, a_{2} \cdots a_{N}) = \left[\sum_{k=1}^{N} a_{k} s(t_{0} - (k-1)T)\right]^{2} - \lambda \sum_{k=1}^{N} \sum_{l=1}^{N} a_{k} a_{l} R_{n}(kT - lT)$$
  
$$\frac{\partial M}{\partial a_{i}} = 0 \implies \frac{\partial M}{\partial a_{i}} = 2 \left[\sum_{k=1}^{N} a_{k} s(t_{0} - (k-1)T)\right] s(t_{0} - (i-1)T) - 2\lambda \sum_{k=1}^{N} a_{k} R_{n}(kT - iT) = 0$$
  
**Because:** 
$$\sum_{k=1}^{N} a_{k} s(t_{0} - (k-1)T) = s_{0}(t_{0})$$
  
$$i = 1, 2 \cdots N$$

Furthermore, let:  $\lambda = s_0(t_0)$   $\longrightarrow_{15}$   $s(t_0 - (i-1)T) = \sum_{k=1}^N a_k R_n(kT - iT)$ 

We define:  

$$s_{i} = s(t_{0} - (i - 1)T)$$

$$r_{ik} = R_{n}(kT - iT)$$

$$i = 1, 2 \dots N$$

$$s(t_{0} - (i - 1)T) = \sum_{k=1}^{N} a_{k}R_{n}(kT - iT)$$

$$s = Ra$$

$$s = \begin{bmatrix} s_{1} \\ s_{2} \\ \vdots \\ s_{N} \end{bmatrix}$$

$$R = \begin{bmatrix} r_{11} & r_{12} & \cdots & r_{1N} \\ r_{21} & r_{22} & \cdots & r_{2N} \\ \vdots & \vdots & \vdots \\ r_{N1} & r_{N2} & \cdots & r_{NN} \end{bmatrix}$$

$$a = \begin{bmatrix} a_{1} \\ a_{2} \\ \vdots \\ a_{N} \end{bmatrix}$$

The coefficients for transversal matched filter are given by

$$a = R^{-1}s$$

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•matched filter is a linear filter that maximizes the instantaneous output signal power for a given input signal waveshape

•For the case of white noise, the impulse response of the matched filter is  $h(t) = Cs(t_0 - t)$ 

•The matched filter can be realized in many forms, Such as the integrate—and—dump, the correlator, and the transversal filter

